

# LED Driver with Average-Mode Constant Current Control



### **General Description**

The FP7122 is an average current mode control LED driver IC operating in a constant off-time mode. FP7122 does not produce a peak-to-average error, and therefore greatly improves accuracy, line and load regulation of the LED current without any need for loop compensation or high-side current sensing. The output LED current accuracy is ±3%.

The FP7122 can be powered from an 8.0 - 100V supply. A PWM & linear dimming input is provided that accepts an external control TTL compatible signal. The output current can be programmed by an internal 250mV reference. or controlled externally through a 0.2 – 1.5V dimming input.

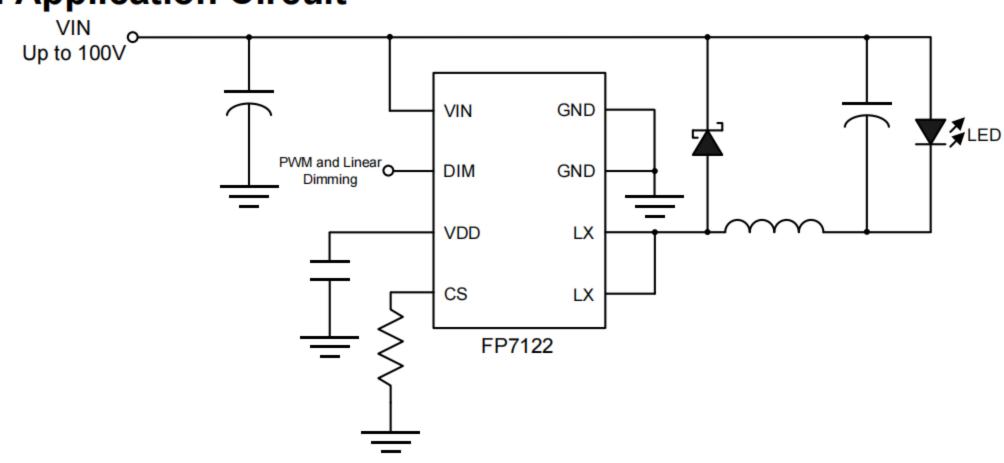
#### **Features**

- > Fast Average Current Control
- ➤ Internal 8 to 100V Linear Regulator
- ➤ Linear and PWM Dimming Capability
- Output Short Circuit Protection with Skip Mode
- ➤ Requires Few External Components for Operation
- Internal 100V/6A N-MOSFET

# **Applications**

- DC/DC or AC/DC LED Driver Applications
- Back Lighting of Flat Panel Displays
- General Purpose Constant Current Source
- Signage and Decorative LED Lighting
- Chargers

# **Typical Application Circuit**

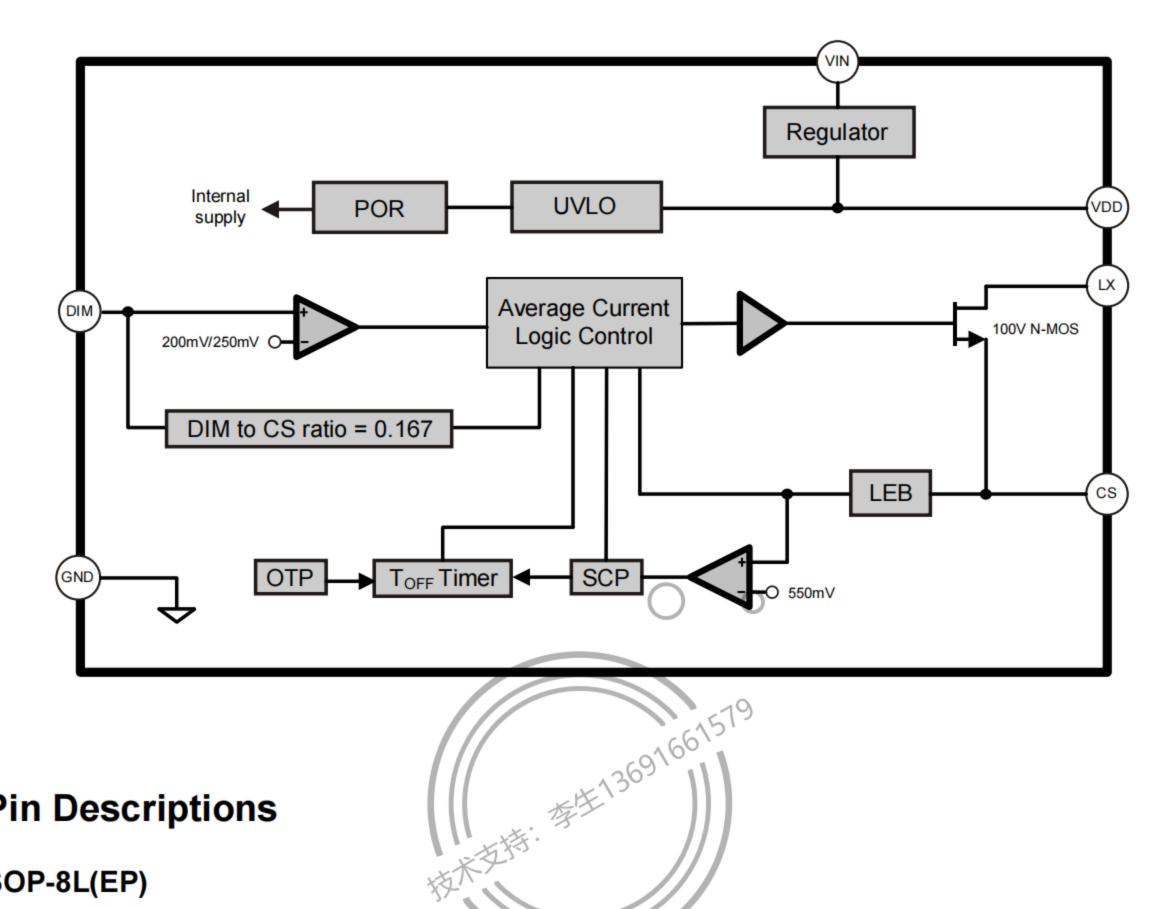


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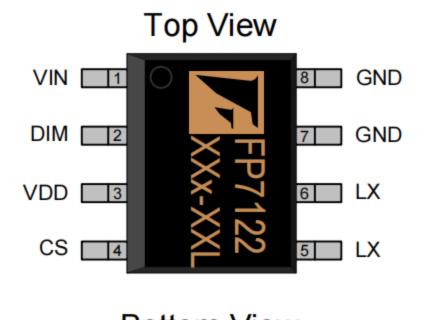


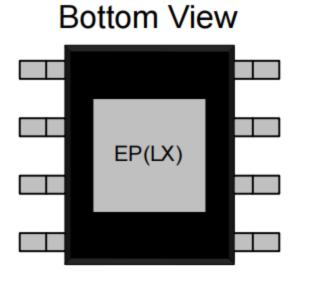
## **Function Block Diagram**



## **Pin Descriptions**

SOP-8L(EP)





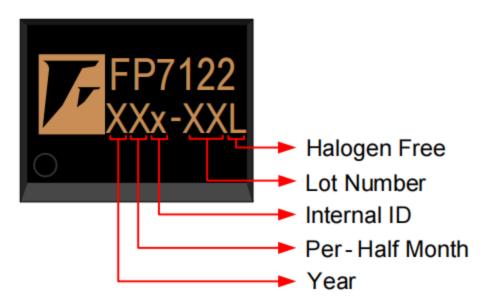
Name	No.	1/0	Description
VIN	1	1	This pin is the input of an 8 - 100V linear regulator.
DIM	2	I	This pin is the linear and PWM dimming input of the IC.
VDD	3	1	This is the power supply pin for all internal circuits.
CS	4	ı	This pin is the current sense pin used to sense the FET current by means of an external sense resistor.
LX	5	0	Power switch
LX	6	0	Power switch
GND	7	Р	Ground return for all internal circuitry.
GND	8	Р	Ground return for all internal circuitry.
EP	9	0	Exposed PAD. Must connect to LX.

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## **Marking Information**



Halogen Free: Halogen free product indicatorLot Number: Wafer lot number's last two digits

For Example  $\longrightarrow$  Lot: 123456  $\longrightarrow$  XXx-56L

Internal ID: Internal Identification Code

Per-Half Month: Production period indicator in half month time unit

For Example :  $A \rightarrow First Half Month of January$ 

B → Second Half Month of January

C → First Half Month of February

D → Second Half Month of February

Year: Production year's last digit

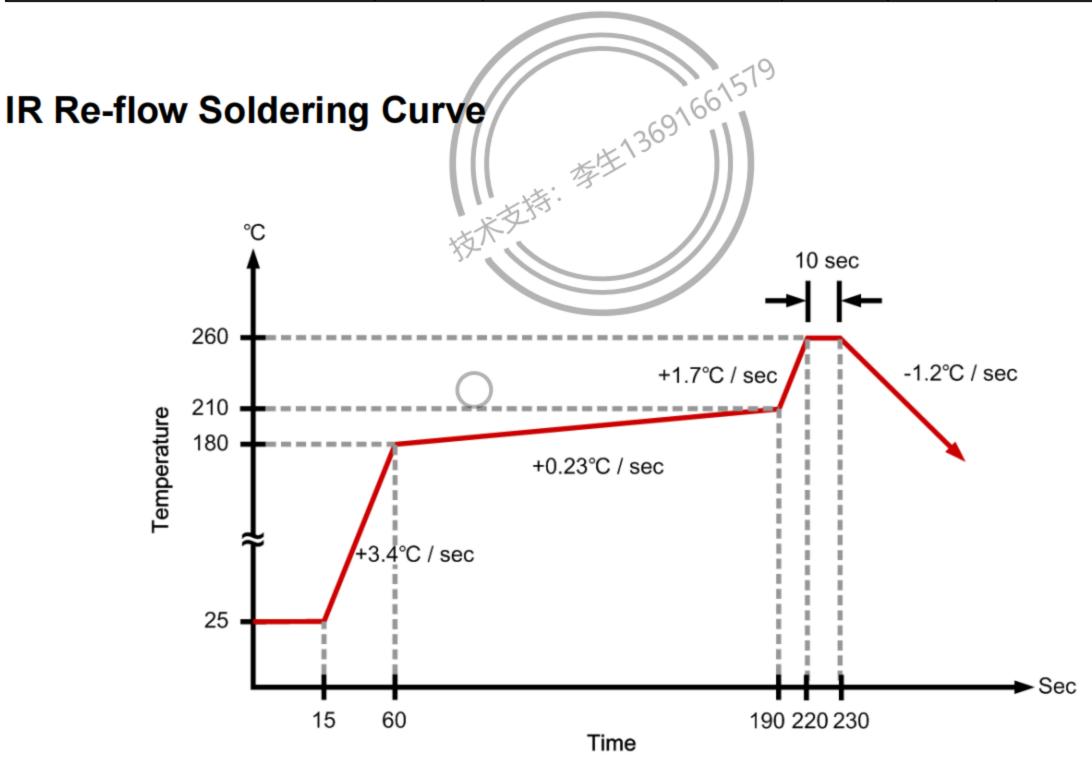


**Ordering Information** 

Part Number	Ambient Operating Temperature	Junction Operating Temperature	Package	MOQ	Description
FP7122XR-G1	-25°C ~ +85°C	-25°C ~ +125°C	SOP-8L(EP)	2500 EA	Tape & Reel

**Absolute Maximum Ratings** 

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
Dower Cumply Voltage	V <sub>IN</sub>	V <sub>IN</sub> to GND			100	٧
Power Supply Voltage	V <sub>DD</sub>	V <sub>DD</sub> to GND			8.5	٧
CS, DIM			-0.3		V <sub>DD</sub> -0.3V	V
Internal MOS V <sub>DS</sub> Breakdown Voltage	V <sub>DS</sub>		100			V
Allowable Power Dissipation	P <sub>D</sub>	SOP-8L(EP) T <sub>A</sub> ≦+25°C			1500	mW
Junction to Ambient Thermal Resistance	θЈА				60	°C / W
Operating Temperature			-25		+85	Ô
Storage Temperature	Ts	SOP-8L(EP)	-40		+150	°C
SOP-8L Lead Temperature		(soldering, 10 sec)			+260	°C



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**Recommended Operating Conditions** 

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
Supply Voltage	V <sub>IN</sub>		8.5		100	V
Ambient Operating Temperature			-25		+85	°C
Junction Operating Temperature			-25		+125	°C

## **DC Electrical Characteristics** ( $V_{IN}$ =12V, $T_A$ = 25 $^{\circ}$ C, unless otherwise noted)

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Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit	
Internal Regulator							
Internally regulated voltage	$V_{DD}$	VIN = 8V, $I_{DD(ext)}$ = 0, 500pF at GATE; DIM= VDD	7.25	7.5	7.75	٧	
Line regulation of VDD	$\Delta V_{\text{DD,line}}$	$VIN = 8 - 100V, I_{DD(ext)} = 0$	0	-	1.0	<b>V</b>	
Load regulation of V <sub>DD</sub>	$\Delta V_{\text{DD,load}}$	$I_{DD(ext)} = 0 - 0.6mA$ , DIM = VDD	0		100	mV	
Shutdown current	I <sub>SD</sub>	V <sub>DIM</sub> =0V		0.4		mA	
V <sub>DD</sub> undervoltage lockout threshold	UVLO	V <sub>DD</sub> rising		6.3		V	
V <sub>DD</sub> undervoltage lockout hysteresis	ΔUVLO	V <sub>DD</sub> falling		500		mV	
PWM Dimming							
Pin DIM input low voltage	V <sub>EN(lo)</sub>	V <sub>IN</sub> = 8 - 100V			0.1	V	
Pin DIM input high voltage	V <sub>EN(hi)</sub>	V <sub>IN</sub> = 8 - 100V	1.6			V	
Average Current Sense Logic	Average Current Sense Logic						
Current sense reference voltage	V <sub>CS</sub>		243	250	257	mV	
DIM-to-CS voltage ratio	A <sub>V(DIM)</sub>		0.164	0.167	0.17		
DIM-to-CS voltage offset	AV <sub>DIM</sub> (OFFSET)	Offset = $V_{CS} - A_{V(DIM)} \cdot V_{DIM}$ $V_{DIM}=1.2V$	0		10	mV	
CS threshold temp regulation					5	mV	
DIM input voltage, shutdown	V <sub>DIM(OFF)</sub>			200		mV	
DIM input voltage, enable	$_{\Delta}V_{\text{DIM(ON)}}$			250		mV	
Current sense blanking interval	T <sub>BLANK</sub>		150		350	ns	
Minimum steady-state duty cycle	T <sub>ON(min)</sub>	CS=V <sub>CS</sub> + 30mV			1000	ns	
Short Circuit Protection							
Hiccup threshold voltage	V <sub>CS</sub>		495	550	605	mV	
Short circuit hiccup time	T <sub>HICCUP</sub>		450	550	650	us	
Minimum on-time (short circuit)	T <sub>ON(min)</sub>	CS=V <sub>DD</sub>			600	ns	

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Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit	
T <sub>OFF</sub> Timer							
Minimum off time	T <sub>OFF(MIM)</sub>			0.6		us	
Maximum off time	T <sub>OFF(MAX)</sub>			50		us	
Maximum on time	T <sub>ON(MAX)</sub>			60		us	
Thermal Protection							
Thermal protection				130		°C	
Thermal shutdown				150		°C	
Internal MOSFET							
Drain-Source Breakdown Voltage	BV <sub>DSS</sub>	V <sub>GS</sub> =0V, I <sub>D</sub> =250uA	100			V	
Drain-Source Leakage Current	I <sub>DSS</sub>	V <sub>DS</sub> =80V, V <sub>GS</sub> =0V,T <sub>J</sub> =25°C			1	uA	
Static Drain-Source On-Resistance <sup>2</sup>	R <sub>DS(ON)</sub>	I <sub>D</sub> =1A			100	mΩ	

Notes:

$$T_{OFF} = \frac{0.6 \times L \times I_{LED(AVG)}}{V_{LED(MAX)}}$$





### **Function Description**

#### **Input Voltage Regulator**

The FP7122 can be powered directly from its VIN pin and can work from 8.5 - 100VDC at its VIN pin. When a voltage is applied at the VIN pin, the FP7122 maintains a constant 7.5V at the VDD pin. This voltage is used to power the IC and any external resistor dividers needed to control the IC. The VDD pin must be bypassed by a low ESR capacitor to provide a low impedance path for the high frequency current of the internal driver.

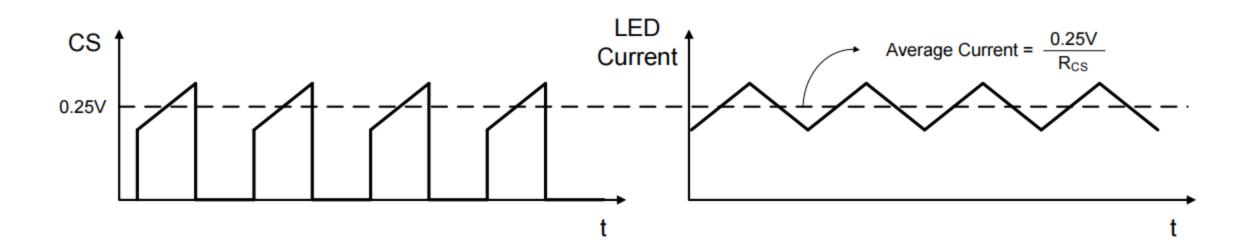
The FP7122 can also be operated by supplying a voltage at the VDD pin greater than the internally regulated voltage. This will turn off the internal linear regulator of the IC and the FP7122 will operate directly off the voltage supplied at the VDD pin. Please note that this external voltage at the VDD pin should not exceed 8.5V.

#### **Average Current Control**

The LED current is detected using a sense resistor at the CS pin. The feedback operates in a fast open-loop mode. No compensation is required. When the voltage at the DIM input  $V_{DIM} \ge 1.5V$ , output current is programmed simply as:

Otherwise:

$$I_{LED}(A) = \frac{V_{DIM}(V) \times 0.167}{R_{CS}(\Omega)}$$

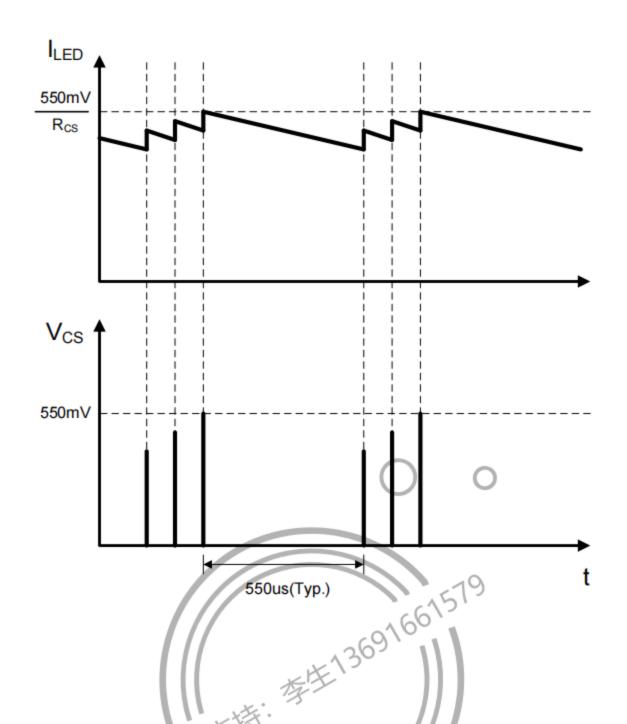


The above equations are only valid for continuous conduction of the output inductor. It is a good practice to design the inductor such that the switching ripple current in it is 25% of its average peak-to-peak, full load, DC current.



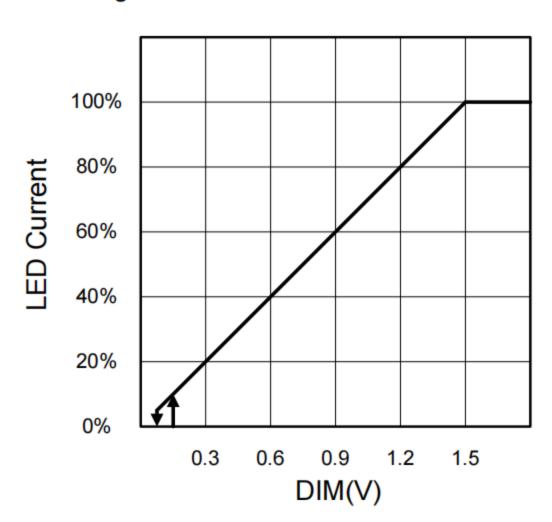
#### **Output Short Circuit Protection**

The short circuit protection comparator trips when the voltage at CS exceeds 0.55V. When this occurs, the GATE off-time  $T_{HICCUP}$  = 550µs is generated to prevent stair-casing of the inductor current and potentially its saturation due to insufficient output voltage.



#### **Linear Dimming**

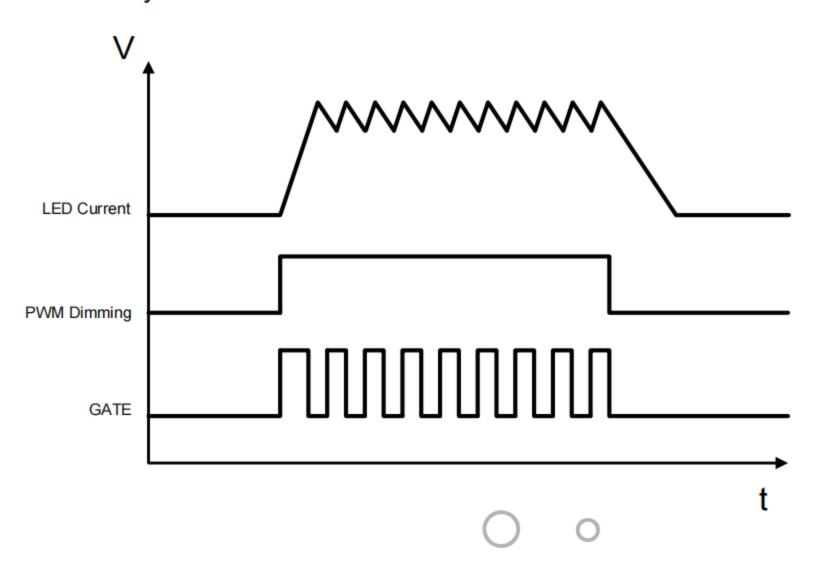
When the voltage at DIM falls below 1.5V, the internal 250mV reference to the constant-current feedback becomes overridden by  $V_{\text{DIM}}$  • 0.185. As long as the current in the inductor remains continuous, the LED current is given by the equation above. However, when  $V_{\text{DIM}}$  falls below 200mV, the internal MOSFET output becomes disabled. The internal MOSFET signal recovers, when  $V_{\text{DIM}}$  exceeds 250mV. This is required in some applications to be able to shut the LED lamp off with the same signal input that controls the brightness.





#### **PWM Dimming**

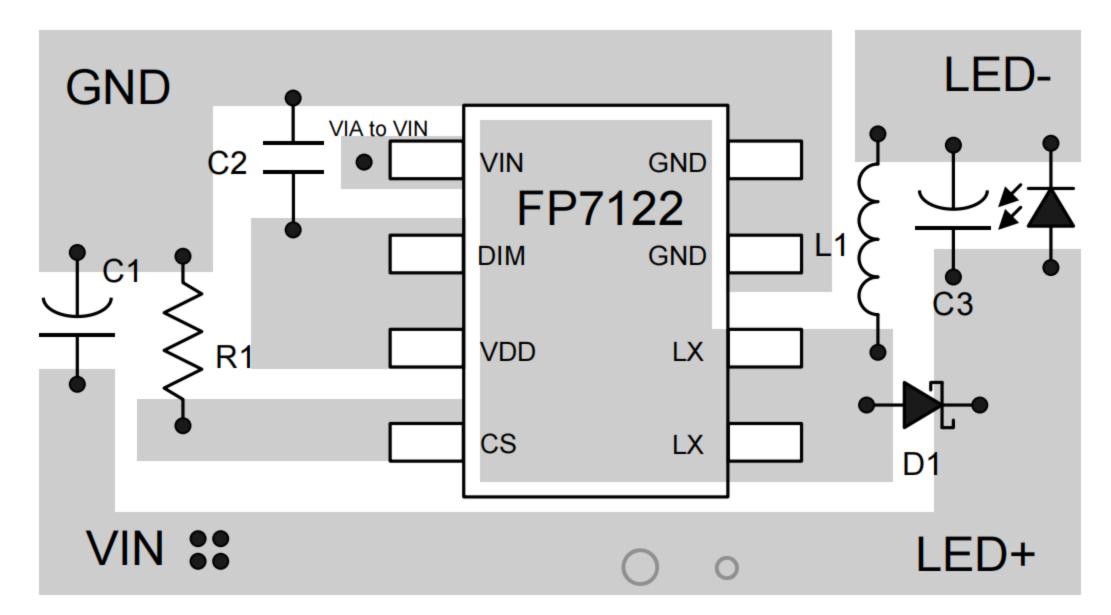
Due to the fast open-loop response of the average-current control loop of the FP7122, its PWM dimming performance nearly matches that of the FP7171.



The rising and falling edges are limited by the current slew rate in the inductor. The first switching cycle is terminated upon reaching the 250 mV ( $V_{\text{DIM}} \cdot 0.167$ ) level at CS. The circuit is further reaching its steady-state within 1 switching cycles regardless of the switching frequency.



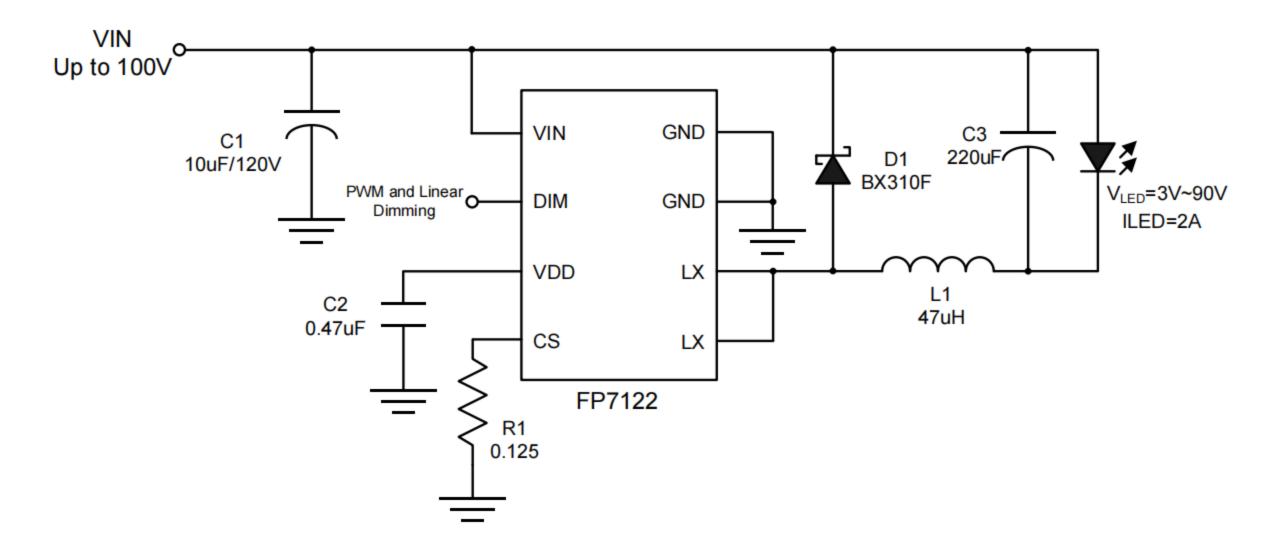
# **Suggested Layout**







# **Typical Application Circuit**





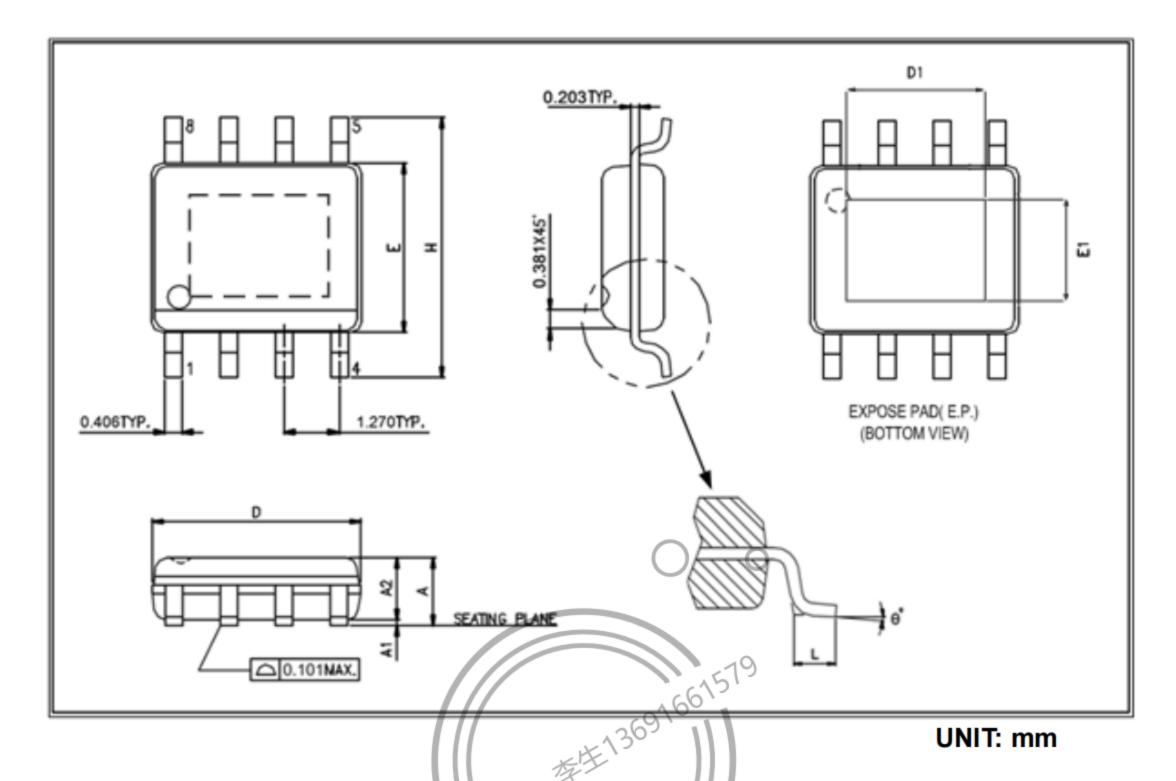
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# Package Outline

### SOP-8L (EP)



Symbols	Min. (mm)	Max. (mm)
Α	1.30	1.70
A1	0	0.15
A2	1.25	1.55
D	4.70	5.10
E	3.80	4.00
Н	5.80	6.20
L	0.40	1.27

### Exposed PAD Dimensions:

Symbols	Min. (mm)	Max. (mm)
D1	2.60	3.45
E1	1.90	2.56

#### Note:

- 1. Package dimensions are in compliance with JEDEC outline: MS-012 AA.
- 2. Dimension "D" does not include molding flash, protrusions or gate burrs.
- 3. Dimension "E" does not include inter-lead flash or protrusions.

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